

Improved Adaptive Impedance Matching for RF Front-End Systems of Wireless Transceivers

Mohammad Alibakhshikenari^{1*}, Bal S. Virdee², Pancham Shukla², Chan H. See^{3,4}, Raed A. Abd-Alhameed⁵, Francisco Falcone⁶, and Ernesto Limiti¹

¹ Electronic Engineering Department, University of Rome “Tor Vergata”, Via del Politecnico 1, 00133, Rome, ITALY

² London Metropolitan University, Center for Communications Technology, School of Computing & Digital Media, London N7 8DB, UK

³ School of Engineering & the Built Environment, Edinburgh Napier University, Merchiston Campus, 10 Colinton Road, Edinburgh, EH10 5DT, UK

⁴ School of Engineering, University of Bolton, Deane Road, Bolton, BL3 5AB, UK

⁵ Faculty of Engineering and Informatics, University of Bradford, Bradford, West Yorkshire, BD7 1DP, UK

⁶ Electric and Electronic Engineering Department, Universidad Pública de Navarra, SPAIN

*alibakhshikenari@ing.uniroma2.it

Abstract- In this paper an automatic adaptive antenna impedance tuning algorithm is presented that is based on quantum inspired genetic optimization technique. The proposed automatic quantum genetic algorithm (AQGA) is used to find the optimum solution for a low-pass passive T -impedance matching LC -network inserted between an RF transceiver and its antenna. Results of the AQGA tuning method are presented for applications across 1.4 to 5 GHz (satellite services, LTE networks, radar systems, and WiFi bands). Compared to existing genetic algorithm-based tuning techniques the proposed algorithm converges much faster to provide a solution. At 1.4, 2.3, 3.4, 4.0, and 5.0 GHz bands the proposed AQGA is on average 75%, 49.2%, 64.9%, 54.7%, and 52.5% faster than conventional genetic algorithms, respectively. The results reveal the proposed AQGA is feasible for real-time application in RF-front-end systems.

Keywords- Antenna impedance matching, automatic quantum genetic algorithm (AQGA), RF-front-end systems, T -type impedance matching LC -network, wireless transceiver-circuits.

I. INTRODUCTION

Antennas provide an interface with the propagating medium and therefore are essential components in wireless communication systems. There is a great demand for a single antenna that can operate efficiently over wide frequency band to accommodate software-defined multi-standard functionality [1]. Conventional wideband antennas for wireless systems are designed to perform satisfactorily, i.e. non-optimum, as the feed-point impedance cannot be matched properly over its operating band. Hence, at the transmitter the efficiency of the high power-amplifier is sub-optimum or at the receiver the low-noise amplifier performance is sub-optimum. To circumvent this issue an impedance matching network needs to be inserted between the antenna and transceiver. Impedance of antennas however can vary significantly with frequency as well as with operating conditions that can result in great impedance variations caused by various factors such as the proximity of a cellular phone to the user's ear [2]. The use of a single impedance matching network with fixed LC component values is therefore unsuitable for realizing accurate matching over a wide frequency range. To realize optimum performance in terms of power and efficiency, it is therefore necessary to utilize an automatically tuneable impedance matching network that can dynamically control the antenna's impedance, frequency band or operating conditions [3–21]. As yet automatically tuneable impedance matching networks are unsuitable for integration in wireless systems because tuning takes too long for practical applications.

Existing impedance tuning algorithms employ a step-by-step methodology where the passive tuning network is modified progressively until the required impedance match goal is realized [3, 22]. Such gradient-based algorithms are computationally intensive because arrival at a solution involves calculation of matrix inversion that can often converge to localised optima [19, 23]. Other algorithms investigated to date to circumvent such issues include (i) genetic algorithms (GA) that use non-gradient and global evolution optimising approaches [3, 24], and (ii) quantum genetic algorithms (QGA) that exploit the laws of quantum mechanics in order to perform efficient computation. Genetic algorithms are essentially adaptive search algorithms based on the evolutionary ideas by Charles Darwin of natural selection and genetics. These search algorithms operate on a set of elements, referred to as population, that evolves by means of crossover and mutation, towards a maximum of the fitness function. QGA have shown to be computationally efficient at solving problems like factorization [25] or searching in an unstructured database [26].

Taking advantage of the quantum to efficiently speed-up classical computation, the QGA outperforms the conventional genetic algorithms (CGA) when the fitness function is varying between genetic iterations. QGA updates the colony by combining quantum bit (qubit) coding with binary coding. It is a kind of efficient parallel algorithm. Qubits are exploited not only to represent the population, but also to perform fitness evaluation and

selection. QGA uses the whole population at each genetic step, and in this sense, it can be considered to be a ‘global search’ algorithm.

Conventional genetic algorithms have been investigated for tuning the impedance of antennas [19-20, 24] but its use has been limited because of issues with obtaining an optimum goal and being slow. Although automatic genetic algorithm (AGA) is an attractive solution for software-defined transceivers it too faces the challenge of achieving an optimum solution quickly as the input impedance of the matching network is a non-linear function of load impedance. Therefore, a more efficient tuning algorithm is needed. Proposed in this paper is impedance tuning of antennas based on automatic quantum genetic algorithm (AQGA) that has been shown to be computationally efficient compared to CGA [26-31] where the fitness function varies continuously. The proposed AQGA is shown to arrive at an optimum impedance matching solution significantly quicker compared to conventional genetic algorithms. In fact, on average at 1.4 GHz band it is 75% faster; at 2.3 GHz band it is 49.2% faster; at 3.4 GHz band it is 64.9% faster; at 4.0 GHz band it is 54.7% faster; and at 5.0 GHz band it is 52.5% faster.

II. AUTOMATIC ANTENNA TUNING SYSTEM (AATS)

The proposed automatic antenna tuning system is designed to conjugately match the impedance of the antenna with the transmitter’s impedance across the operating frequency band of the wireless system. The AATS, shown in Fig.1, comprises (i) a tuneable matching network that generates the desired impedance transformation; (ii) an impedance sensor to measure the VSWR at the matching network input; and (iii) a control unit that modifies the impedance of the matching network to the desired impedance based on feedback from the impedance sensor by using the proposed tuning algorithm. The matching system can respond to variations in the antenna impedance affected by changes in its operating conditions.

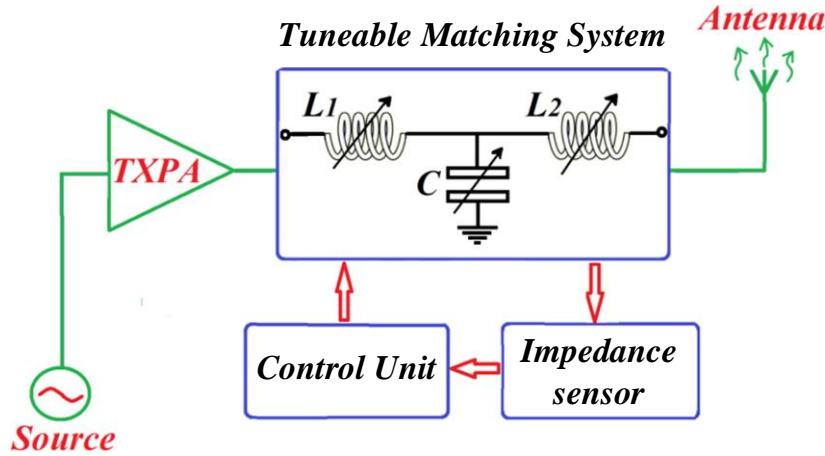


Fig.1. Block diagram of the automatic antenna tuning system (AATS).

III. IMPEDANCE MATCHING LC-NETWORK

As the antenna impedance is a function of the frequency and operational conditions, therefore the impedance matching network should be capable of being dynamically adaptable. The matching process involves suitably controlling the parameters of the matching network. The low-pass π -type and T -type impedance matching LC-networks are attractive configurations as they enable the use of a broad range of load impedances and harmonic-rejection characteristics [4], [31].

The proposed low-pass passive T -type impedance matching LC-network, which is located between the transmitter and antenna, is shown in Fig.2, where Z_{source} is the source impedance of the transmitter, and Z_{Load} represents the load impedance. Impedances Z_A , Z_B and Z_{input} are annotated in Fig.2 at each transformation stage from the load to the source. To realize impedance matching Z_{input} must equal the real part of the source impedance, R_{source} . Hence, the maximum available power transfer to the load and Voltage Standing Wave Ratio (VSWR) can easily be determined from Fig.2 as following

$$Z_A = R_{Load} + j(X_{Load} + \omega L_2) \quad (1)$$

$$Z_B = \frac{R_{Load} + j[X_A - \omega C(X_A^2 + R_{Load}^2)]}{(1 + \omega C X_A)^2 + (\omega C R_{Load})^2} \quad (2)$$

Where

$$X_A = X_{Load} + \omega L_2$$

$$Z_{input} = Z_B + j\omega L_1 \quad (3)$$

$$\text{or} \quad Z_{input} = A + jB \quad (4)$$

$$\text{Where} \quad A = \frac{R}{(1+\omega CX_A)^2 + (\omega CR_{Load})^2} \quad \text{and} \quad B = \omega L_1 + \frac{X_A - \omega C(X_A^2 + R_{Load}^2)}{(1+\omega CX_A)^2 + (\omega CR_{Load})^2}$$

Reflection-coefficient (Γ_{input}) at input of the LC-network is defined by

$$\Gamma_{input} = \frac{Z_{input} - Z_{source}}{Z_{input} + Z_{source}} \quad (5)$$

$$\text{and} \quad VSWR = \frac{1 + |\Gamma_{input}|}{1 - |\Gamma_{input}|} \quad (6)$$

Maximum available power from source is

$$P_{max_{av}} = \frac{|V_{source}|^2}{4Re(Z_{source})} \quad (7)$$

Power delivered to the LC-network can be derived and is given by

$$P_{input} = \frac{|V_{source}|^2}{8Z_o} \frac{|1 - \Gamma_{source}|^2}{|1 - \Gamma_{source}\Gamma_{input}|^2} (1 - |\Gamma_{input}|^2) \quad (8)$$

Where Γ_{source} is reflection-coefficient at the source, and Z_o is the characteristic impedance.

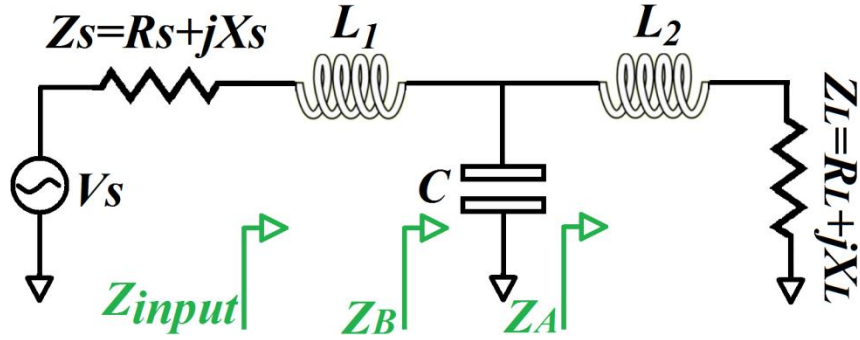
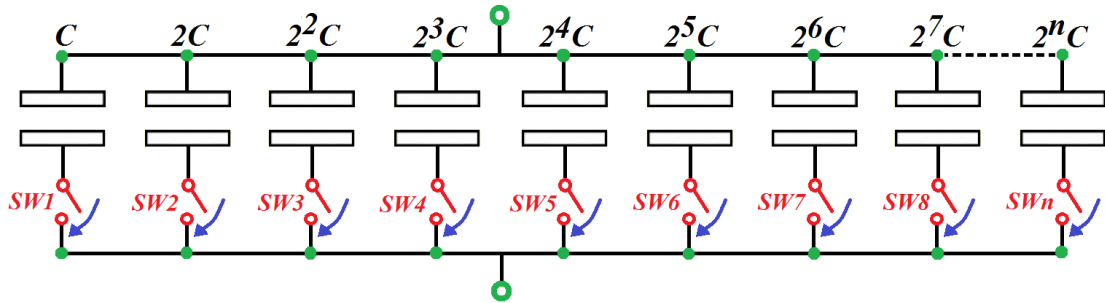
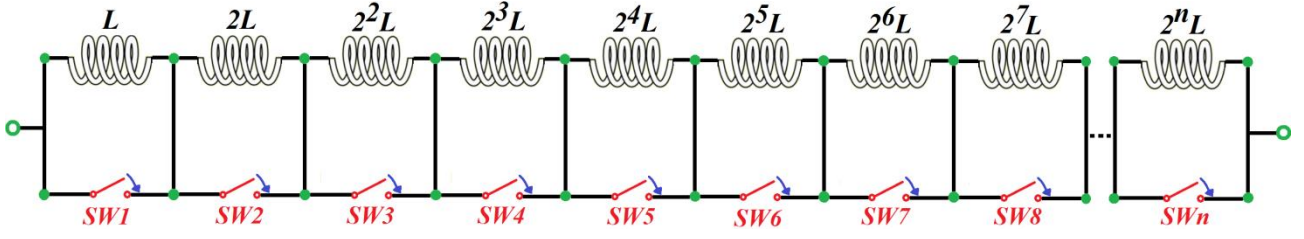


Fig.2. T-type impedance matching LC-network.

In the proposed T-type impedance matching LC-network the magnitude and range of the reactive components should cover the impedances necessary to match the antenna across the wireless systems operating bandwidth. The reactive components should be controllable to accommodate variations in the antenna's operating conditions. To implement the tuning requirement an array of digital switches of fixed value reactive component are deployed, as shown in Fig.3. This arrangement facilitates automatic adjustment of the matching network to meet the matching condition (ideally $\Gamma_{source} = 0$ and $VSWR = 1$) using a tuning algorithm.



(a) Array of digital switched capacitors



(b) Array of digital switched inductors

Fig. 3 Digital switched LC-network arrays.

IV. PROPOSED AUTOMATIC QUANTUM GENETIC ALGORITHM

Fast tuning algorithm is required for antenna impedance networks especially with changing loads and operational conditions. In the case where many network combinations are possible it is important to have a tuning algorithm that can reduce the number of networks possible. This can be achieved by generating a look-up table of matching networks for various frequencies in the operating range. This approach will facilitate the control system to rapidly select a suitable tuning network as a function of frequency [23]. The disadvantage of this approach is it's not able to respond quickly to changes in the antenna's impedance as a function of time without repeating the whole process again.

Genetic tuning algorithms, however, based on iterations converge on the best impedance networks [24, 26, 27]. Such algorithms must initially perform many iterations to arrive at a satisfactory impedance solution with no need for explicit rules. But with continued operation such algorithms achieve matching with fewer iterations. The disadvantage of GA is it requires significant computational time.

It has been shown that AQGA is an efficient algorithm [24, 28–31] as it's (i) less prone to being trapped in a localised optimum solution; (ii) requiring less iteration; and (iii) it converges more rapidly to a final solution. It is for these reasons the tuning mechanism chosen here is based on AQGA.

A. Representation

AQGA is a quantum inspired genetic algorithm that is based on quantum computation and conventional GA. Its states are a superposition of qubits [24, 28]. A qubit can assume state $|0\rangle$ or $|1\rangle$, or any superposition of the dual states defined by $|\delta\rangle = i|0\rangle + j|1\rangle$, where parameters i and j represent complex numbers indicating the probability-amplitudes of the respective states. For a system of n qubits, the system can exhibit 2^n states concurrently. Its representation is given by [4, 31]

$$\begin{bmatrix} i_1 & i_2 & i_3 & i_4 & i_5 & \dots & i_n \\ j_1 & j_2 & j_3 & j_4 & j_5 & \dots & j_n \end{bmatrix} \quad (9)$$

where

$$|i_x|^2 + |j_x|^2 = 1, \quad x = 1, 2, 3, 4, 5, \dots, n \quad (10)$$

One qubit chromosome in Eqn.(9) can represent all possible states in the primary stages of evolution, whereas 2^n chromosomes are required in a classical system.

For the T -type impedance matching LC-network in Fig.2, the components (L_1 , L_2 , and C) are coded in the form of Eqn.(9). The chromosome $[L_1, L_2, C]$ comprises 30-bit qubits, where 10-bit qubits represent each component. The quantum algorithm's challenge is to determine the minimum list of M items.

B. Structure of AQGA

The structure of the AQGA for modifying the T -type impedance matching LC-network is thus [4, 31]:

- i) Initialize the binary instants and the quantum population (do this for each generation)
- ii) Compute the fitness of each entity P_f^x
- iii) Categorize the entities related to the fitness amounts and store the best chromosome
- iv) Apply the quantum genetic operators on $P_{Quantum}^x$ and update qubit chromosome applying rotation matrix

The initial component magnitudes of the T -type impedance matching LC-network are achieved from the previous section and are adjusted by random numbers where the magnitudes of the probability amplitudes are selected in a random fashion from the intervals of $L_1, L_2 \in \{10^{-12}, 10^{-6}\}$ and $C \in \{10^{-14}, 10^{-7}\}$. Constituents of the quantum population are represented as

$$P_{Quantum}^x = [Q_1^x, Q_2^x, Q_3^x, Q_4^x, Q_5^x, \dots, Q_m^x] \quad (11)$$

where

$$Q_y^x = \{i_{y,l}^x, j_{y,l}^x\}^R \quad (12)$$

Q_y^x is the l^{th} qubit size in the x^{th} generation of the y^{th} constituent in the quantum population (P_Y^x) and m is the population size. P_Y^x is generated from qubit chromosome $P_{Quantum}^x$, and represented as

$$P_Y^x = [Y_1^x, Y_2^x, Y_3^x, Y_4^x, Y_5^x, \dots, Y_m^x] \quad (13)$$

where Y_y^x is the x^{th} bit of the chromosome.

The quantum matrix is transformed into a binary matrix in the measurement operation. As done in other quantum systems a single solution is extracted from the quantum matrix while preserving all other configurations. The magnitude of the qubit is determined according to its probability pairs $|i_{y,l}^x|^2$ and $|j_{y,l}^x|^2$. Population of binary entities is built from the quantum population $P_{Quantum}^x$. Each qubit is observed for any destruction in the qubit chromosome. Diversity in population is achieved by generating a random number. $|1\rangle$ state is recorded whenever the magnitude of the random number is bigger than the corresponding chromosome probability amplitude. If the magnitude of the random number is smaller compared to the corresponding chromosome probability amplitude $|0\rangle$ state will be observed.

In the evaluation phase the fitness of the current and the best chromosome is given by $f(CH_{\beta}^{\alpha})$ and $f(CH_{best})$, respectively, where CH_{best} is the α^{th} bit of the best chromosome. $f(CH_{best})$ is obtained from

$$f(CH_{best}) = \text{maximum}\{f(CH_1^{\alpha}), f(CH_2^{\alpha}), f(CH_3^{\alpha}), f(CH_4^{\alpha}), f(CH_5^{\alpha}), \dots, f(CH_m^{\alpha})\} \quad (14)$$

For the T -type impedance matching LC -network, the population is evaluated by

$$f(L_1, L_2, C) = \left(1 + |Z_{source} - Z_{input}|^2\right) / |\Gamma_{source}| \quad (15)$$

where Z_{input} is the population's actual input impedance, and Γ_{source} is the source's actual reflection-coefficient. The aim of the tuning algorithm is to determine the values of L_1 , L_2 , and C to meet the matching conditions. This is followed by updating the x^{th} population of qubit chromosomes $P_{Quantum}^x$ by applying quantum rotation.

Qubit chromosome of $P_{Quantum}^x$ with the best fitness is chosen for each binary chromosome P_Y^x . Then qubit chromosomes are sorted according to the values of the fitness in every iteration executed. To avoid convergence to a local maximum, the selection strategy used here was to extract the optimum and part of "not so good" entities. In this way global optimisation is achieved. A uniform quantum crossover operation is applied thereafter to the chosen entities. This is followed by mutating the probability amplitude of each qubit chromosome to generate new entities. Finally, the quantum chromosome is updated by using quantum rotation. The rotation matrix is represented by

$$R(\varphi) = \begin{bmatrix} \sin(\varphi) & \cos(\varphi) \\ \cos(\varphi) & -\sin(\varphi) \end{bmatrix} \quad (16)$$

where φ represents the angle of the rotation. The l^{th} qubit in the x^{th} generation is

$$Q_y^x = \{i_{y,l}^x, j_{y,l}^x\}^R \quad (17)$$

which is updated as

$$\begin{bmatrix} i_{y,l}^x \\ j_{y,l}^x \end{bmatrix}_{new} = T(\varphi_{y,l}^x) \begin{bmatrix} i_{y,l}^x \\ j_{y,l}^x \end{bmatrix}_{old} \quad (18)$$

where $\varphi_{y,l}^x$ is the corresponding rotation angle of Q_y^x .

$$Q_{y,new}^x = \{i_{y,l}^x, j_{y,l}^x\}_{new}^R \quad (19)$$

$$Q_{y,old}^x = \{i_{y,l}^x, j_{y,l}^x\}_{old}^R \quad (20)$$

$$\varphi_{y,l}^x = d_{y,l}^x |\varphi_{y,l}^x| \quad (21)$$

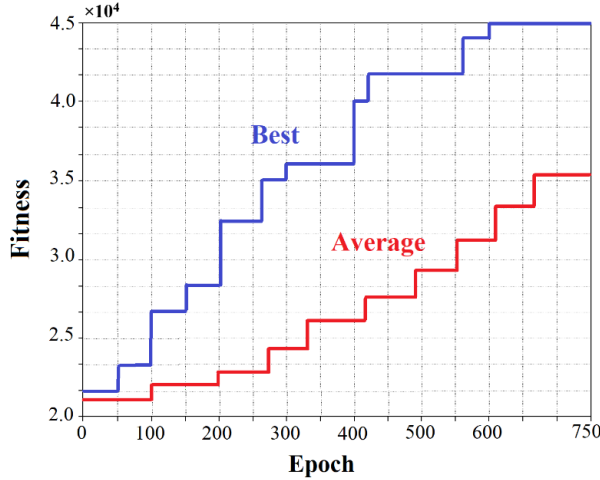
where $|\varphi_{y,l}^x|$ is its amplitude and $d_{y,l}^x$ is its sign. $d_{y,l}^x$ is dependent on $\Psi = i_{y,l}^x \cdot j_{y,l}^x$. Table 1 shows the relation between the fitness condition $d_{y,l}^x$ and " Ψ ", where the mark sensing $\Psi > 0$ stands $Re(\Psi) \cdot Im(\Psi) > 0$, and $\Psi < 0$ denotes for $Re(\Psi) \cdot Im(\Psi) < 0$. In case of $\Psi = 0$, i.e. $i_{y,l}^x = 0$ or $j_{y,l}^x = 0$, there are different $d_{y,l}^x$ amounts, whose details are presented in Table 1. The symbolism "*" illustrates arbitrary state, symbol "1" in the columns demonstrates for the clockwise rotation of $d_{y,l}^x$, and "0" in the columns corresponded to $d_{y,l}^x$ indicates that rotation is not accomplished. The rotation operation helps AQGA to converge rapidly to find a solution.

TABLE 1. Strategy of the Quantum Rotation

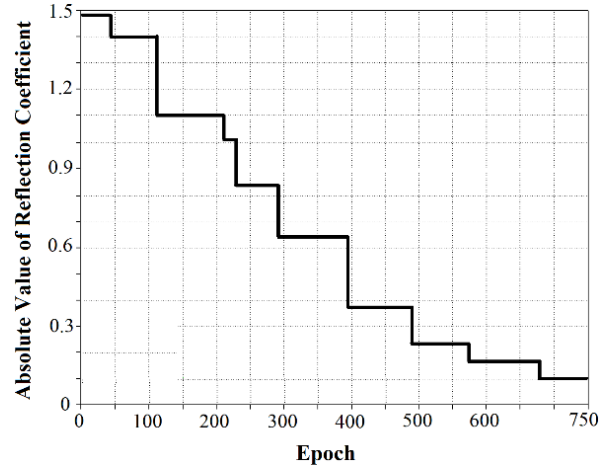
CH_β^α	CH_{best}	$f(CH_\beta^\alpha) \geq f(CH_{best})$	$ \varphi_{y,l}^x $	$d_{y,l}^x$			
				$\Psi > 0$	$\Psi < 0$	$\Psi = 0$	
						$i_{y,l}^x = 0$	$j_{y,l}^x = 0$
1	1	*	*	0	0	0	0
1	0	$T (true)$	0.08π	-1	1	*	0
1	0	T	0.04π	1	-1	0	*
0	1	$F (false)$	0.08π	1	-1	0	*
0	1	F	0.04π	-1	1	*	0
0	0	*	*	0	0	0	0

V. RESULTS AND DISCUSSIONS

The proposed automatic quantum GA tuning approach is compared with the conventional GA methodology to verify its effectiveness. It was therefore necessary to extensively simulate both tuning approaches. The tuning times of the algorithms was assessed over 400 simulations to determine the algorithm's speed in arriving at the optimal solution. The training diagrams of the AQGA tuning approach are shown in Fig.4 for the following conditions: $Z_{source}=50+j30 \Omega$, $Z_{load}=75+j50 \Omega$, signal frequency of 5 GHz (WiFi band), and signal amplitude of 1 V. The results shown are a function of size of genetic population, probability of mutation, cross-probability, specific mutation, cross and selection operations. The AQGA tuning was executed many times for various parameter magnitudes. The best and average fitness results in Fig.4 corresponds to a population size of 100 entities, a mutation probability of 0.5% and a cross-probability of 12.4%. The best entities are chosen by the probability of 85% and the remainders are chosen by the probability of 15%. The algorithm terminated after 750 generations and the optimum solution achieved at this point is applied, which results in $L_1=5.62 nH$, $L_2=3.45 nH$, and $C=2.18 pF$. The quantum elapsed time is 7.25s. Fig.4(b) shows the corresponding reflection-coefficient variations with training epochs. Table 2 also gives the different load impedance results at 5 GHz (WiFi band).



(a) Best and average fitness



(b) Reflection-coefficient

Fig.4. Training plots of the AQGA tuning for $Z_{source}=50+j30 \Omega$, $Z_{load}=75+j50 \Omega$, signal frequency of 5 GHz (WiFi band), & signal amplitude of 1 V.

In another example, AQGA tuning is applied to source impedance of $Z_{source}=50+j30 \Omega$, load impedance of $Z_{load}=75+j50 \Omega$, source signal frequency of 3.4 GHz (radar systems), and signal amplitude of 1 V. Computational results in Fig.5 are for a population size of 100 entities, a mutation probability of 0.6% and a cross-probability of 13.6%. Best entities are chosen by the probability of 87% and the remainders are selected by the probability of 13%. The algorithm terminates after 750 generations and the optimum goal achieved gives $L_1=7.11 nH$, $L_2=4.25 nH$, and $C=3.05 pF$. The quantum elapsed time is 7.02s. Table 2 also gives the results for different load impedances at 3.4 GHz (radar systems).

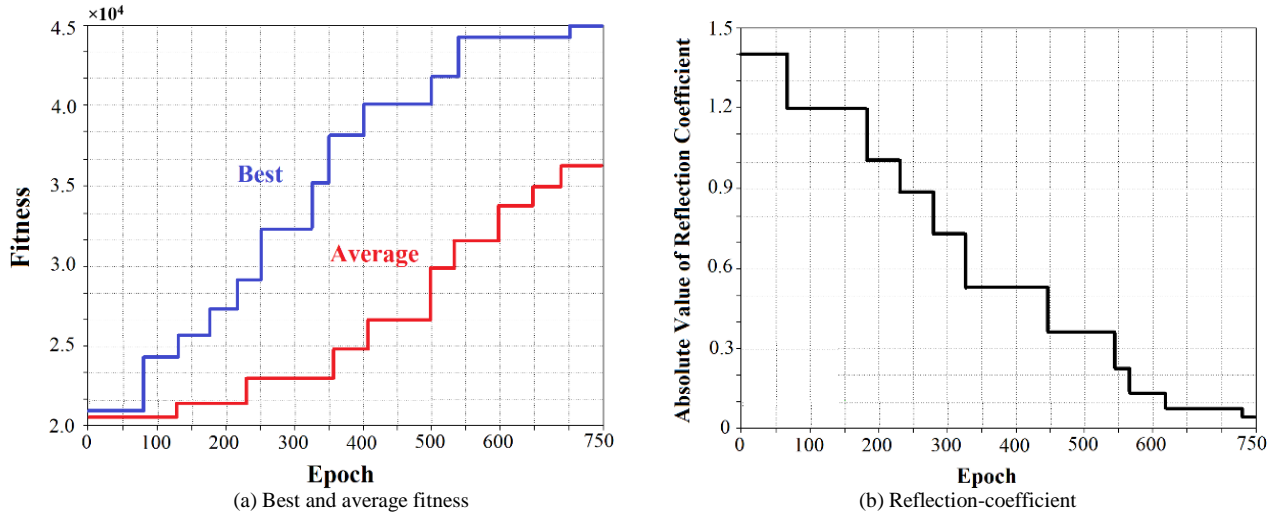


Fig.5. Training curves of the AQGA tuning for $Z_{source}=50+j30 \Omega$, $Z_{Load}=75+j50 \Omega$, signal frequency of 3.4 GHz (radar systems), & signal amplitude of 1 V.

Statistical results of 400 executions for various load impedances applying the proposed AQGA are presented in Table 2 for different frequency bands, i.e. 1.4 GHz (military and satellite services), 2.3 GHz (LTE networks), 3.4 GHz (radar systems), 4.0 GHz (satellite earth stations), and 5.0 GHz (WiFi band), where the relative error in input impedance is defined as

$$E_{Relative} = |Z_{input} - Z_{source}|/Z_{source} \quad (22)$$

where Z_{input} is the input impedance determined from the tuning algorithm, and $Z_{source}=50+j30 \Omega$ is the source impedance to be matched. As can be seen from Table 2, the errors of the input impedance and VSWR obtained are negligible. Summarized in Table 2 is the results of the training epoch, relative error, and the reflection coefficient for various load impedances at various frequencies.

The conventional GA tuning approach was also carried out for comparison purposes using the same impedance matching network at various frequencies of 1.4 GHz (military and satellite services), 2.3 GHz (LTE networks), 3.4 GHz (radar systems), 4.0 GHz (satellite earth stations), and 5.0 GHz (WiFi band). The numerical results are given in Table 2. It is evident from the table that the proposed AQGA tuning method arrives at the required component value using significantly less iterations in comparison to the CGA tuning method.

TABLE 2. Results of 400 Runs for Different Load Impedances at Various Frequencies Using the Proposed AQGA and CGA Tuning Approaches

Frequency Band		$Z_{Load} (\Omega)$	Train Epochs	Relative Error (%)	VSWR
CGA	$f = 1.4$ GHz (1350.5-1515 MHz) (satellite services)	$25 + j50$	332542	2.45×10^{-4}	1.2492
		$25 - j50$	525278	2.97×10^{-5}	1.3548
		$50 + j75$	576985	6.55×10^{-3}	1.2254
		$50 - j75$	548481	4.92×10^{-5}	1.1458
		$75 + j100$	188094	1.58×10^{-4}	1.2094
		$75 - j100$	179792	2.14×10^{-3}	1.1983
AQGA	$f = 1.4$ GHz (1350.5-1515 MHz) (satellite services)	$25 + j50$	270	9.86×10^{-6}	1.1368
		$25 - j50$	328	3.71×10^{-6}	1.0557
		$50 + j75$	360	2.07×10^{-3}	1.1521
		$50 - j75$	345	1.21×10^{-4}	1.0692
		$75 + j100$	290	1.70×10^{-5}	1.0958
		$75 - j100$	265	1.55×10^{-4}	1.1410
CGA	$f = 2.3$ GHz (2350-2390 MHz) (LTE networks)	$25 + j50$	31347	2.26×10^{-3}	1.4059
		$25 - j50$	21942	3.98×10^{-3}	1.2602
		$50 + j75$	20563	1.62×10^{-3}	1.3058
		$50 - j75$	30725	2.81×10^{-3}	1.4069
		$75 + j100$	19497	4.20×10^{-4}	1.3693
		$75 - j100$	22051	5.95×10^{-4}	1.1690
AQGA	$f = 2.3$ GHz (2350-2390 MHz) (LTE networks)	$25 + j50$	256	6.83×10^{-4}	1.2275
		$25 - j50$	291	3.12×10^{-5}	1.1508
		$50 + j75$	275	1.03×10^{-4}	1.1821
		$50 - j75$	213	3.55×10^{-4}	1.2551
		$75 + j100$	272	4.07×10^{-4}	1.1953

		75 - j100	219	6.73×10^{-5}	1.0565
CGA	$f = 3.4$ GHz (3410-3600 MHz) (radar systems)	25 + j50	28653	3.24×10^{-3}	1.3429
		25 - j50	32961	4.01×10^{-2}	1.2630
		50 + j75	25061	3.98×10^{-4}	1.2918
		50 - j75	19239	1.53×10^{-3}	1.7392
		75 + j100	24716	3.65×10^{-3}	1.2530
		75 - j100	15630	2.64×10^{-4}	1.3709
AQGA	$f = 3.4$ GHz (3410-3600 MHz) (radar systems)	25 + j50	189	2.02×10^{-4}	1.2093
		25 - j50	267	6.16×10^{-3}	1.1196
		50 + j75	345	3.86×10^{-3}	1.1532
		50 - j75	220	5.37×10^{-4}	1.0514
		75 + j100	198	1.17×10^{-3}	1.1090
		75 - j100	254	4.29×10^{-5}	1.0961
CGA	$f = 4.0$ GHz (3700-4200 MHz) (satellite earth stations)	25 + j50	234975	6.32×10^{-3}	1.3095
		25 - j50	286531	4.61×10^{-2}	1.2783
		50 + j75	295514	2.84×10^{-4}	1.2590
		50 - j75	301860	1.90×10^{-3}	1.4061
		75 + j100	287461	3.66×10^{-3}	1.3068
		75 - j100	310639	2.38×10^{-4}	1.2831
AQGA	$f = 4.0$ GHz (3700-4200 MHz) (satellite earth stations)	25 + j50	429	7.51×10^{-4}	1.1009
		25 - j50	401	1.39×10^{-3}	1.1891
		50 + j75	385	1.01×10^{-4}	1.2029
		50 - j75	393	4.42×10^{-4}	1.3007
		75 + j100	438	2.58×10^{-3}	1.2117
		75 - j100	419	5.98×10^{-5}	1.1503
CGA	$f = 5.0$ GHz (5350-5925 MHz) (WiFi band)	25 + j50	28537	2.47×10^{-3}	1.4280
		25 - j50	19620	3.85×10^{-3}	1.3891
		50 + j75	15352	1.29×10^{-4}	1.3631
		50 - j75	22385	5.38×10^{-2}	1.3284
		75 + j100	29716	4.50×10^{-3}	1.2801
		75 - j100	18630	2.48×10^{-3}	1.4429
AQGA	$f = 5.0$ GHz (5350-5925 MHz) (WiFi band)	25 + j50	247	1.24×10^{-4}	1.1503
		25 - j50	209	4.92×10^{-4}	1.2745
		50 + j75	278	1.01×10^{-5}	1.1645
		50 - j75	192	1.58×10^{-2}	1.1716
		75 + j100	273	3.54×10^{-4}	1.1550
		75 - j100	203	4.97×10^{-4}	1.3509

The AQGA and CGA tuning algorithms were executed on a Pentium (R) 7 CPU 16 GHz PC. The longest, shortest, and average time for 400 runs for both algorithms are given in Table 3. The proposed AQGA approach obtains the solution much more rapidly compared with the CGA approach.

TABLE 3. Comparison of Computation Times of AQGA and CGA Approaches (Unit: Seconds)

Frequency Band	Algorithm	Shortest	Longest	Average	Compared to CGA the proposed AQGA is faster on average by
1.4 GHz (military & satellite services)	CGA	5.3528	49.3091	26.1073	75%
1.4 GHz (military & satellite services)	Proposed AQGA	3.2850	10.8324	6.5049	
2.3 GHz (LTE networks)	CGA	2.9841	52.1520	27.6302	49.2%
2.3 GHz (LTE networks)	Proposed AQGA	1.7536	28.9173	14.0371	
3.4 GHz (radar systems)	CGA	6.8453	65.7329	35.2961	64.9%
3.4 GHz (radar systems)	Proposed AQGA	4.2740	21.6510	12.3874	
4.0 GHz (satellite earth stations)	CGA	4.9560	42.0845	22.0351	54.7%
4.0 GHz (satellite earth stations)	Proposed AQGA	2.0352	17.3842	9.9835	
5.0 GHz (Wi-Fi band)	CGA	3.1271	31.0653	17.9350	52.5%
5.0 GHz (Wi-Fi band)	Proposed AQGA	1.0735	15.0252	8.5193	

The proposed approach of single frequency tuning can be applied for tuning a given RF band. This involves first tuning the matching network to the center frequency of the operating band of the system. Component values obtained for other frequencies in the given band are then used to compute the VSWR values. As an example, let's consider the WiFi band (5350 – 5925 MHz) with a bandwidth of 575 MHz. The T -type impedance matching network is tuned at the WiFi's centre frequency of 5.63 GHz for source impedance of $Z_{source}=50+j30\ \Omega$, load impedance of $Z_{Load}=75+j50\ \Omega$, and signal amplitude of 1 V. The algorithm provides impedance matching component values of $L_1=6.50\ nH$, $L_2=4.36\ nH$ and $C=2.95\ pF$. The VSWR as a function of frequency is shown in Fig. 6 by applying these component values. The VSWR varies in a V-shape response across its band between 1.15 and 1.35. This shows the VSWR performance is acceptable over the whole bandwidth from 5350 – 5925 MHz after the tuning algorithm converges.

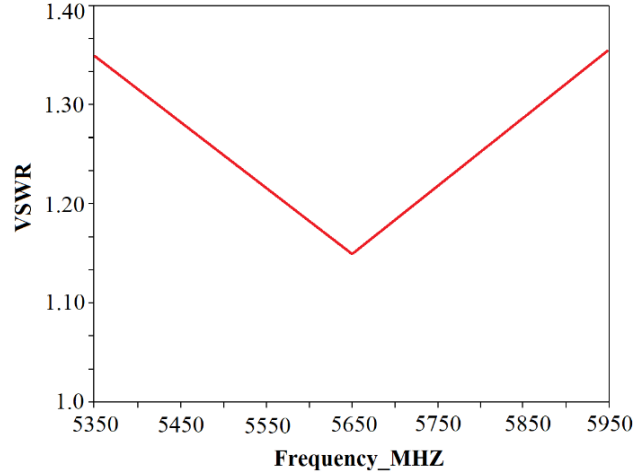


Fig.6. Voltage standing wave ratio (VSWR) over spectrum for WiFi band from 5350 – 5925 MHz.

Fig. 7 show the VSWR frequency response for a wider RF band defined between 1400 to 3400 MHz, i.e. ISM band in which WLAN, Wi-Fi and Bluetooth operate. In this case the VSWR falls in the range between 1.05 and 1.25. This result shows the effectiveness of the proposed algorithm.

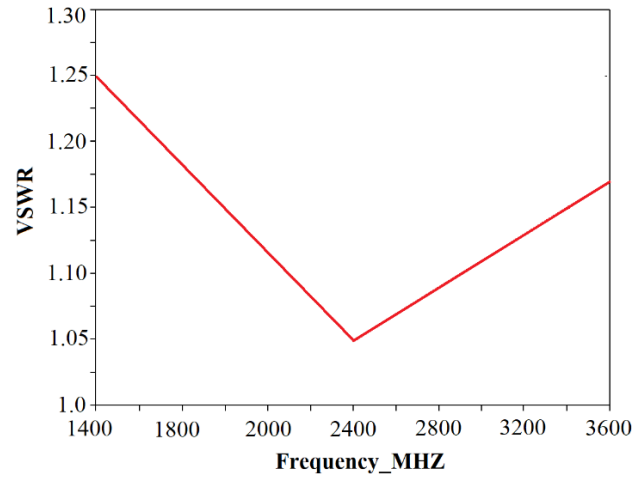


Fig.7. Voltage standing wave ratio (VSWR) over spectrum for ISM band (1400 – 3400 MHz).

In the case for much wider RF bands, the tuning to the band centre may not give the accuracy required. It is therefore proposed to use multiple frequencies for tuning to realize one set of optimum component values that cover the specified frequency band for a pre-set error.

VI. CONCLUSIONS

Effectiveness of the proposed automatic quantum genetic algorithm (AQGA) is demonstrated against the conventional genetic algorithm (CGA) to adaptively modify the impedance of the matching network to conjugately match the impedance of the antenna with the transmitter's impedance across the operating frequency band of the wireless system. Optimum matching solution is reached in a significantly shorter time compared to CGA. The proposed algorithm was applied in several scenarios to tune the T -type impedance matching LC -network at different frequency bands. The results presented verify the effectiveness of the proposed AQGA tuning approach for real-time adaptive antenna tuning.

AUTHORS CONTRIBUTIONS

Conceptualization, M.A., B.S.V., P.S., C.H.S., R.A.A.-A., F.F., and E.L.; methodology, M.A., B.S.V., F.F., E.L.; software, M.A., P.S., and C.H.S.; validation, M.A., B.S.V., P.S., and C.H.S.; formal analysis, M.A., F.F., and E.L.; investigation, M.A., P.S., C.H.S., and R.A.A.-A.; resources, M.A., B.S.V., P.S., C.H.S., R.A.A.-A., F.F., and E.L.; data curation, M.A., B.S.V., C.H.S., and R.A.A.-A.; writing—original draft preparation, M.A.; writing—review and editing, M.A., B.S.V., P.S., C.H.S., R.A.A.-A., F.F., and E.L.; visualization, M.A., B.S.V., P.S., C.H.S., R.A.A.-A., F.F., and E.L.; supervision, E.L.; project administration, R.A.A.-A., F.F., and E.L.; funding acquisition, R.A.A.-A., F.F., E.L..

ACKNOWLEDGMENTS

This work is partially supported by RTI2018-095499-B-C31, Funded by Ministerio de Ciencia, Innovación y Universidades, Gobierno de España (MCIU/AEI/FEDER,UE), and innovation programme under grant agreement H2020-MSCA-ITN-2016 SECRET-722424 and the financial support from the UK Engineering and Physical Sciences Research Council (EPSRC) under grant EP/E022936/1.

REFERENCES

- [1] Saidatul, N.A., Soh, P.J., Sun, Y., Lauder, D., Azremi, A.A.H.: ‘Multiband fractal PIFA (planar inverted F antenna) for mobile phones’, *Proc. IEEE Int. Symp. Wireless Communications Systems*, York, UK, September 2010.
- [2] Boyle, K.R., Yuan, Y., Lighthart, L.P.: ‘Analysis of mobile phone antenna impedance variations with user proximity’, *IEEE Trans. Antennas Propag.*, 2007, 55(2), pp. 364–372.
- [3] Sun, Y., Moritz, J., Zhu, X.: ‘Adaptive impedance matching and antenna tuning for green software defined and cognitive radio’. Invited Paper in special session on Signal Processing for Software Defined and Cognitive Radio, 54th IEEE Midwest Symp. Circuits and Systems, Seoul, Korea, August 2011.
- [4] Tan, Y., Yi, R., Sun, Y.: ‘Wideband tuning of impedance matching networks using hierarchical genetic algorithms for multistandard mobile communications’, *J. Comput.*, 2012, 7(2), pp. 356–361.
- [5] Mileusnic, M., Petrovic, P., Todorovic, J.: ‘Design and implementation of fast antenna tuners for HF radio systems’, *ICICS ‘97*, Singapore, 9–12 September 1997, 3B3.1, pp. 1722–1725.
- [6] Moritz, J.R., Lauder, D.M., Sun, Y.: ‘Measurement of HF antenna impedances’, *IEE Colloquium on Frequency Selection and Management Techniques for HF Communications*, Ref No. 1999/017, London, UK, 29–30 March, 1999, pp. 141–147.
- [7] Ida, I., Takada, J., Toda, T., Oishi, Y.: ‘An adaptive impedance matching system and its application to mobile antennas’, *Proc. IEEE TENCON*, 2004, 3, pp. 543–546.
- [8] De Mingo, J., Valdovinos, A., Crespo, A., Navarro, D., Carcia, P.: ‘An RF electronically controlled impedance tuning network design and its application to an antenna input impedance automatic matching systems’, *IEEE Trans. Microw. Theory Tech.*, 2004, 52(2), pp. 489–497.
- [9] Zolomy, A., Mernyei, F., Erdelyi, J., Pardoen, M., Toth, G.: ‘Automatic antenna tuning for RF transmitter IC applying high Q antenna’, *IEEE Radio Frequency Integrated Circuits Symp.* 2004, 6–8 June, 2004, paper TU4D-4, pp. 501–504.
- [10] Meng, F., Bezooijen, A., Mahmoudi, R.: ‘A mismatch detector for adaptive antenna impedance matching’, *Proc. 36th European Microwave Conf.*, Manchester, UK, September 2006, pp. 1457–1460.
- [11] Oh, S.-H., Song, H., Aberle, J.T., Bakkaloglu, B., Chakrabarti, C.: ‘Automatic antenna tuning unit for software-defined and cognitive radio’, *Wirel. Commun. Mobile Comput.*, 2007, 7, pp. 1103–1115.
- [12] Song, H., Bakkaloglu, B., Aberle, J.T.: ‘A CMOS adaptive-antenna impedance tuning IC operating in the 850 MHz-to-2 GHz band’, *IEEE Int. Solid-State Circuits Conf.*, Session 22, 2009, pp. 18–20.
- [13] Chamseddine, A., Haslett, J.W., Okoniewski, M.: ‘CMOS silicon-on-sapphire tunable matching networks’, *EURASIP Journal on Wireless Communications and Networking*, 2006, article ID 86531, pp. 1–11.
- [14] Sjoblom, P., Sjoland, H.: ‘An adaptive impedance tuning CMOS circuit for ISM 2.4-GHz band’, *IEEE Trans. Circuits Syst.*, 1, 2005, 52(6), pp. 1115–1124.
- [15] Sjoblom, P., Sjoland, H.: ‘Measured CMOS switched high-quality capacitors in a reconfigurable matching network’, *IEEE Trans. CAS-II*, 2007, 54(10), pp. 858–862.
- [16] Firrao, E.L., Annema, A.J., Nauta, B.: ‘An automatic antenna tuning system using only signal amplitudes’, *IEEE TCAS-II*, 2008, 55(9), pp. 833–837.
- [17] Van Bezooijen, A., de Jongh, M.A., Chanlo, C., et al.: ‘A GSM/EDGE/ WCDMA adaptive series-LC matching network using RF-MEMS switches’, *IEEE J. Solid-State Circuits*, 2008, 43(10), pp. 2259–2268.
- [18] Van Bezooijen, A., de Jongh, M.A., van Straten, F., Mahmoudi, R., van Roermund, H.M.: ‘Adaptive impedance-matching techniques for controlling L networks’, *IEEE Trans. Circuits Syst. I, Regul. Pap.*, 2010, 57(2), pp. 495–505.
- [19] Ogawa, K., Takahashi, T., Koyanagi, Y., Ito, K.: ‘Automatic impedance matching of an active helical antenna near a human operator’. *Thirty-third European Microwave Conf.*, 2003, pp. 1271–1274.
- [20] Sun, Y., Lau, W.K.: ‘Automatic impedance matching using Genetic Algorithms’. *Proc. IEE Conf. Antennas and Propagation*, York, UK, August 1999, pp. 31–36.

- [21] S. Soltani, P. Lotfi, and R.D. Murch: 'Design and optimization of multiport pixel antennas', IEEE Transactions on Antennas and Propagation, 66 (4), April 2018, pp. 2049-2054.
- [22] A. M. Elshurafa and E. I. El-Masry, "Tunable matching networks for future MEMS-based transceivers," Proc. IEEE Int. Symp. Circuits and Systems, Vol. IV, pp.457-460, 2004.
- [23] De Mingo, J., Valdovinos, A., Crespo, A., Navarro, D., Carcia, P.: 'An RF electronically controlled impedance tuning network design and its application to an antenna input impedance automatic matching systems', IEEE Trans. Microw. Theory Tech., 2004, 52, (2), pp. 489–497
- [24] Thompson, M., Fidler, J.K.: 'Application of the genetic algorithm and simulated annealing to LC filter tuning', IEE Proc. Circuits Devices Syst., 2001, 148, (4), pp. 177–182.
- [25] Shor, P. W.: 'Algorithms for quantum computation: Discrete logarithms and factoring', Proc. 35th Annu. Symp. Found. Comput. Sci., Los Alamitos, CA, Nov. 1994, pp. 124–134.
- [26] Grover, L. K.: 'Quantum mechanics helps in searching for a needle in a haystack,' Phys. Rev. Lett, 79(2), pp. 325–328, Jul. 1997.
- [27] Han, K.H., Kim, J.H.: 'Quantum-inspired evolutionary algorithm for a class of combinatorial optimization', IEEE Trans. Evol. Comput., 2002, 6(6), pp. 580–592.
- [28] Q. Y. Liu, X. J. Meng, Wang, S. H. et al: 'Design Optimization of Power Transformer Based on Multilevel Genetic Algorithm', Journal of XI'ANJIAOTONG University, 43 (6), pp.113 – 117, June, 2009.
- [29] Malossini, A., Blanzieri, E., Calarco, T.: 'Quantum genetic optimization', IEEE Trans. Evol. Comput., 2008, 12(2), pp. 231–241.
- [30] Narayanan, A., Moore, M.: 'Quantum inspired genetic algorithm'. Proc. Int. Conf. Evolutionary Computation, 1996, pp. 61–66.
- [31] Tan, Y., Sun, Y., and Lauder, D.: 'Automatic impedance matching and antenna tuning using quantum genetic algorithms for wireless and mobile communications,' IET Microw. Antennas Propag., 2013, 7(8), pp. 693–700.